

Signature  
Name (print, please)  
Student No.

**ECE 2300 Circuit Analysis**  
**Summer 2009**  
**July 2, 2009**

**Exam 1**

**DO NOT OPEN THIS EXAM BOOKLET UNTIL  
INSTRUCTED TO DO SO**

*This exam has 10 pages including this cover page. If you are missing any pages,  
raise your hand. You have 90 minutes to complete the exam.*

**Notes**

1. Be sure your name and signature appear above.
2. The quiz is closed-book. You may have a calculator and one 8 ½" x 11" crib sheet.
3. To receive full credit for a problem, you must:
  - Show all work necessary to solve the problem;
  - Define all variables and parameters and label them on circuit diagrams;
  - Use the proper notation for all variables.
  - Show all units explicitly in intermediate and final results;
  - Indicate clearly whether power being calculated is absorbed or delivered;

1. \_\_\_\_\_/25

2. \_\_\_\_\_/25

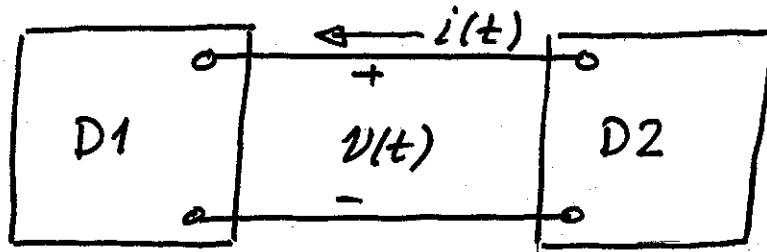
3. \_\_\_\_\_/25

4. \_\_\_\_\_/25

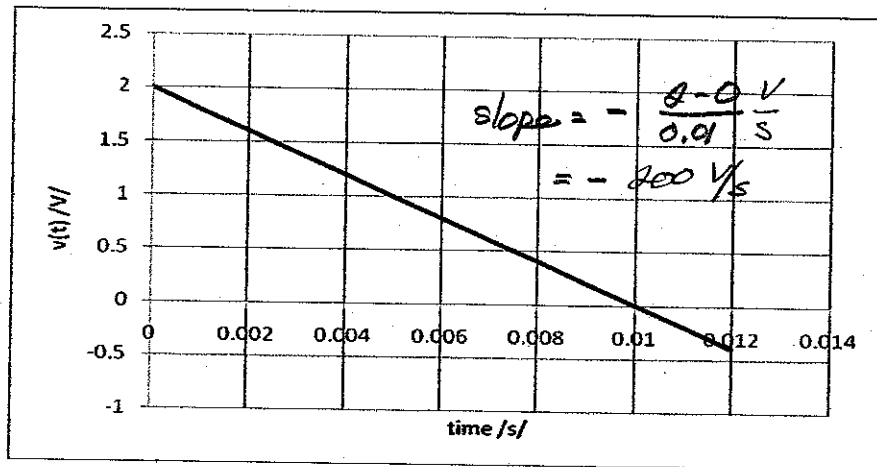
Total \_\_\_\_\_/100

1. (25 points) The box labeled D1 can be modeled using a current source in parallel with a resistance. The box labeled D2 can be modeled using a voltage source in series with a resistance. The voltage  $v(t)$  and current  $i(t)$  labeled in the figure have the time dependence shown in the graphs below the figure.

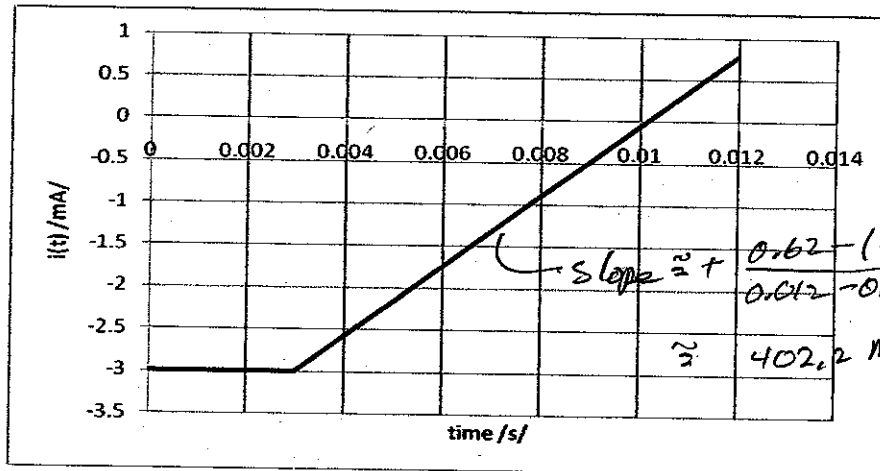
- Find the power being delivered by D2 to D1 at  $t = 10 \text{ ms}$ .
- Find the total energy delivered by D2 to D1 between 0 and 10 ms.



$v(t)$   
→



$i(t)$   
→



Approximate slope:

$$\text{slope} \approx + \frac{0.62 - (-3)}{0.012 - 0.003} \frac{\text{mA}}{\text{s}} \approx 402.2 \text{ mA/s} = 0.402 \text{ A/s}$$

over →

Room for extra work

a) We will need to work out the linear functions representing  $v(t)$  and  $i(t)$ :

$$v(t) = m_v t + b_v \quad b_v = 2 \text{ V} \text{ by inspection}$$

$$m_v = \frac{2-0}{0-0.01} = -200 \text{ V/s}$$

$$\therefore \boxed{v(t) = 2 - 200t \text{ V} \quad t \text{ in } \mu\text{s}}$$

$$\boxed{i(t) = +3 \text{ mA} \quad 0 \leq t \leq 3 \text{ } \mu\text{s}}$$

$$i(t) = m_i t + b_i \quad 3 \leq t \leq 12 \text{ } \mu\text{s}$$

$$-0.0005 = m_i (0.009) + b_i$$

$$-0.0030 = m_i (0.003) + b_i$$

$$\Rightarrow m_i = 0.4167 \text{ A/s} \quad b_i = -4.25 \text{ mA}$$

$$\therefore \boxed{i(t) = 0.4167t - 4.25 \times 10^{-3} \text{ A} \quad t \text{ in } \mu\text{s}}$$

Now it is clear that  $t = 10 \text{ } \mu\text{s} \Rightarrow v(t) = 0$

so

$$\boxed{P_{\text{abs}, DI} = 0 \text{ at } t = 10 \text{ } \mu\text{s}}$$

Room for extra work

$$b) \quad W_{\text{del by } P_2} = \int_0^{0.01 \text{ s}} P_{\text{del by } P_2} dt = + \int_0^{0.01 \text{ s}} v(t) \cdot i(t) dt$$

$$\begin{aligned} v(t) \cdot i(t) &= -0.003(2 - 200t) \text{ W} & \left. \begin{array}{l} 0 \leq t \leq 0.003 \text{ s} \\ \end{array} \right\} \\ &= -0.006 + 0.6t \text{ W} \end{aligned}$$

$$\begin{aligned} v(t) \cdot i(t) &= (-4.25 \times 10^{-3} + 0.4167t) \cdot (2 - 200t) \text{ W} \\ &= -8.34 \times 10^{-3} t^2 + 1.684t - 8.5 \times 10^{-3} \text{ W} & \left. \begin{array}{l} 0.003 \leq t \leq 0.01 \text{ s} \\ \end{array} \right\} \end{aligned}$$

$$\begin{aligned} \therefore W_{\text{del by } P_2} &= \int_0^{0.003 \text{ s}} (-0.006 + 0.6t) dt + \int_{0.003}^{0.01 \text{ s}} (-8.34t^2 + 1.684t - 8.5 \times 10^{-3}) dt \\ &= -1.53 \times 10^{-5} \text{ J} - 9.927 \times 10^{-6} \text{ J} \end{aligned}$$

$$W_{\text{del by } P_2} = -2.523 \times 10^{-5} \text{ J}$$

formulation +6

sign +2

calculations +9

+11

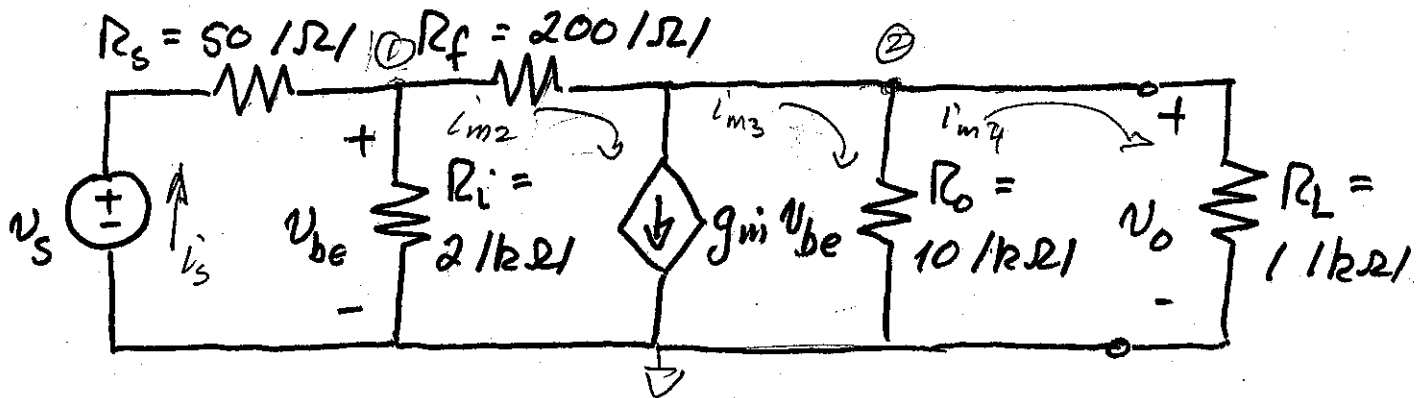
divide limits +4

2. (25 points) The circuit below is a model for what is known as a *transconductance amplifier*. The voltage  $v_s$  represents an input signal; the load at the output is modeled by the resistor  $R_L$ . The parameter  $g_m$  is the transconductance, and has a value of  $150 \text{ /mS/}$ .

a) If  $v_s = 1 \text{ /V/}$ , find  $v_o$ .

b) Find the power delivered to the rest of the circuit by the signal voltage  $v_s$ .

c) Find the power delivered to the load  $R_L$ .



a) Looks like a job for node voltage... we have labeled the reference node as well as the non-ref. nodes ① and ②. We note that these node voltages are already labeled  $v_o$  and  $v_{be}$ . So...

Eqs: +15

$$\textcircled{1} \quad \frac{v_{be} - v_s}{R_s} + \frac{v_{be}}{R_i} + \frac{v_{be} - v_o}{R_f} = 0$$

$$\textcircled{2} \quad \frac{v_o - v_{be}}{R_f} + g_m v_{be} + \frac{v_o}{R_o} + \frac{v_o}{R_L} = 0$$

From here we will substitute numbers to make the going easier. we will also substitute  $v_s = 1 \text{ /V/}$ .

$$\frac{v_{be} - 1}{50} + \frac{v_{be}}{2000} + \frac{v_{be} - v_o}{200} = 0$$

$$\frac{v_o - v_{be}}{200} + 0.150 v_{be} + \frac{v_o}{10000} + \frac{v_o}{1000} = 0$$

Room for extra work

Solving these equations gives

$$V_{be} = 0.1386 \text{ V} \quad V_o = -3.293 \text{ V}$$

a)  $\underline{V_o = -3.293 \text{ V}}$  + 3

b)  $\underline{P_{del \ v_s} = V_s \cdot i_s = 1 \cdot (17.23 \times 10^{-3}) = 17.23 \text{ mW}}$  + 2 sign  
+ 3

$$i_s = \frac{V_s - V_{be}}{50} = \frac{1 - 0.1386}{50} = 17.23 \text{ mA}$$

c)  $\underline{P_{abs \ R_L} = \frac{V_o^2}{R_L} = \frac{(-3.293)^2}{1000} = 10.84 \text{ mW}}$  + 2

Alternative solution using mesh currents:

①  $-V_s + 50 i_s + 2000 (i_s - i_{m2}) = 0$   $V_s = 1 \text{ V}$

②  $200 i_{m2} + 10000 (i_{m3} - i_{m4}) + 2000 (i_{m2} - i_s) = 0$

③  $i_{m2} - i_{m3} = g_m \cdot 2000 (i_s - i_{m2})$   $g_m = 0.150 \text{ S}$

④  $1000 i_{m4} + 10000 (i_{m4} - i_{m3}) = 0$

Soln:  $i_{m4} = -3.29 \times 10^{-3} \text{ A}$

$$\underline{i_s = 1.72 \times 10^{-2} \text{ A}}$$

$$i_{m2} = 1.72 \times 10^{-2} \text{ A}$$

$$i_{m3} = -3.62 \times 10^{-3} \text{ A}$$

$\Rightarrow \underline{V_o = -3.29 \text{ V}}$

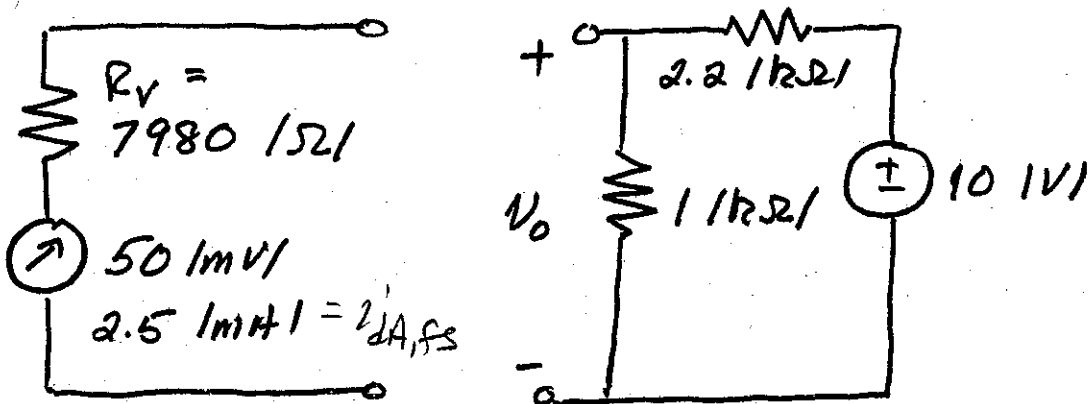
-6- so same answers.

3. (25 points) Depending on the value of  $R_V$ , the voltmeter shown below can be set to full-scale readings of 5, 10, 20, 25, 50, 80, 100, 125, 150, and 200 V/. In the figure,  $R_V$  has been chosen so that the full-scale reading is 20 V/.

- 10 a) If the voltmeter is used on the 20 V/ full scale range to measure the voltage  $v_o$  in the circuit on the right, what is the percent error in the reading? Assume that the "true" voltage  $v_{o,true}$  is the value of  $v_o$  when the meter is not connected to the circuit, and that the percent error  $E$  is given by

$$E = 100 \left( \frac{v_{o,measured} - v_{o,true}}{v_{o,true}} \right)$$

- 15 b) What is the smallest full scale voltage setting that can be used and still ensure that the magnitude of the percent error is no more than 2%?



1) d'Arsonval meter resistance is  $R_{DA} = \frac{50 \mu V}{2.5 mA} = 20 \Omega$

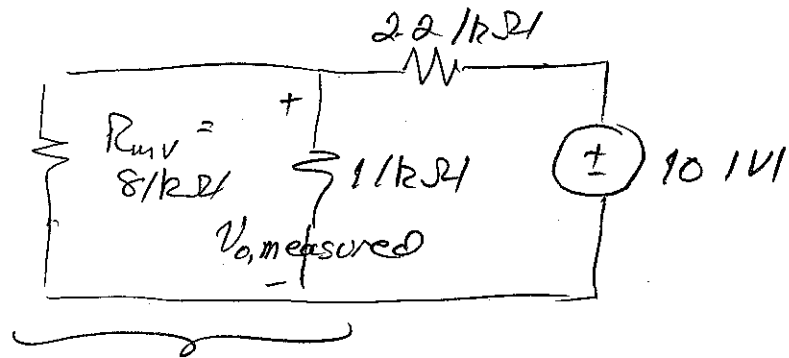
meter resistance  $\Rightarrow R_{mv} = 7980 + 20 = 8000 \Omega$

Thus full scale rating is  $I_{DA,fs} \cdot R_{mv} = 20 V$   
(as stated)

$$v_{o,true} = 10 \cdot \frac{1 k\Omega}{1 k\Omega + 2.2 k\Omega} = 3.125 V$$

Room for Extra Work

with the meter attached, we have



$$R_{eq} = 8k \parallel 1k = 888.9 \Omega$$

$$V_{o,measured} = 10 \cdot \frac{888.9}{888.9 + 2200} = 2.878 \text{ V}$$

$$\text{and } E = \frac{2.878 - 3.125}{3.125} = -2.91\%$$

ii) -2% error means that

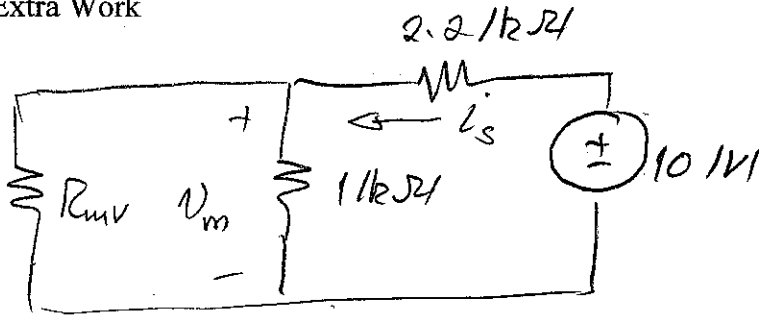
$$\frac{V_{o,measured} - V_{o,true}}{V_{o,true}} = -0.02$$

$$\Rightarrow V_{o,measured} = -0.02 V_{o,true} + V_{o,true} = 3.0625 \text{ V}$$

If that's to be true, then we can find the required  $R_{mv}$ :



Room for Extra Work



$$R_{eq} = R_{mv} \parallel 1k\Omega$$

If  $V_m$  is 3.0625 V, then  $I_s' = \frac{10 - 3.0625}{2200}$

$\Rightarrow I_s' = 3.153 \text{ mA}$ , Thus

$$R_{eq} = \frac{3.0625 \text{ V}}{3.153 \text{ mA}} = 971.2 \Omega$$

So then

$$\frac{R_{mv} \cdot 1k\Omega}{R_{mv} + 1k\Omega} = 971.2 \Omega$$

$$\Rightarrow R_{mv} = \frac{0.19712 \text{ k}\Omega}{1 - 0.9712}$$

$$R_{mv} = \frac{0.19712 \text{ k}\Omega}{1 - 0.9712} = 33.72 \text{ k}\Omega$$

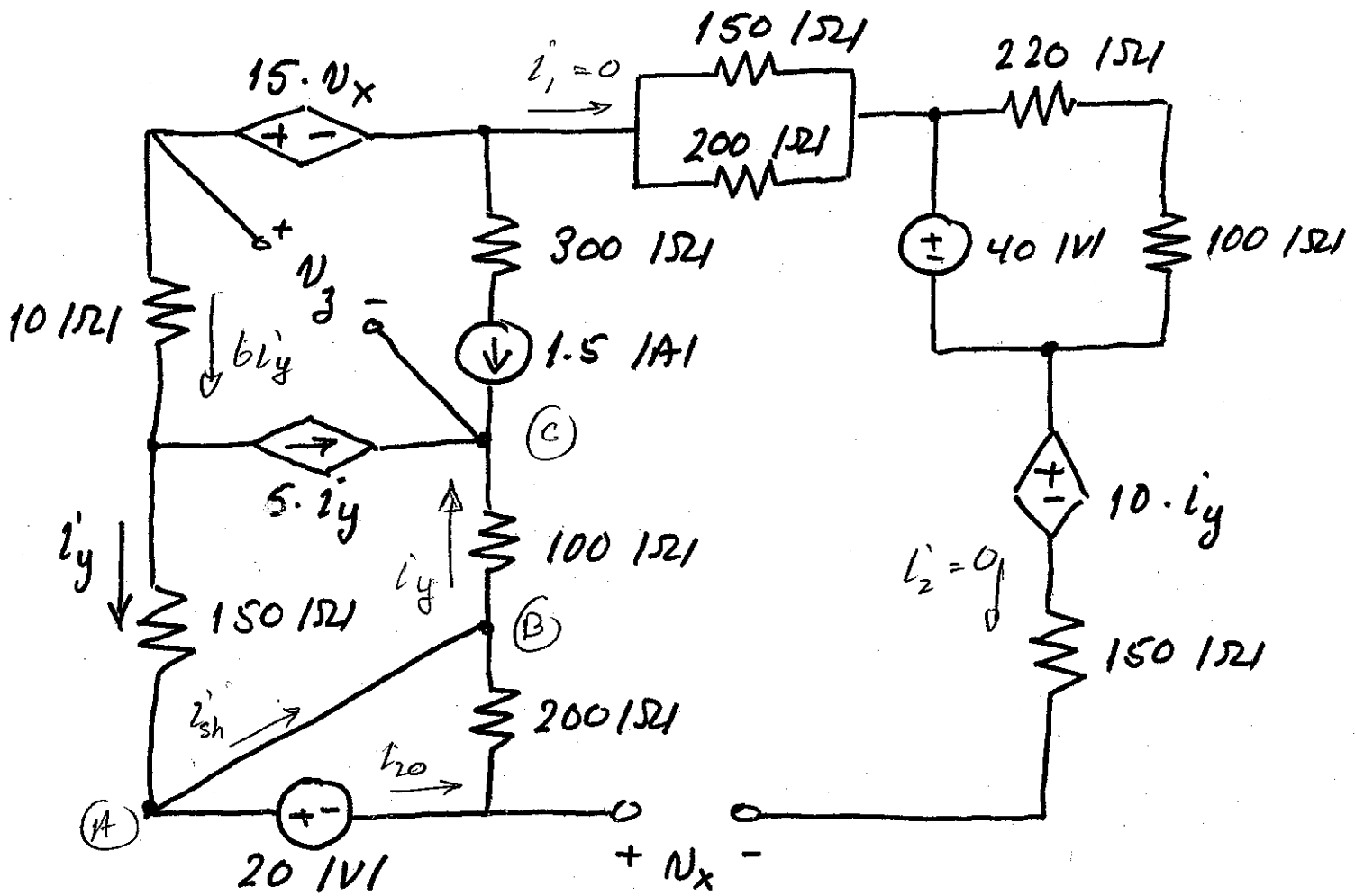
+7

Finally, if the voltmeter has a full scale current of 2.5 mA and a resistance 33.72 kΩ, the full-scale voltage reading is

$$+2 \quad V_{fs} = I_{fs} \cdot R_{mv} = 84.30 \text{ V}$$

This is not available so we go to the next higher range, which is 100 V full-scale.

4. (25 points) For the circuit below, find  $v_z$ .



The current in the  $100 \Omega$  resistor is  $i_y$ , as shown. We can show this as follows:

$$\text{KCL: } i_y = i_{sh} + i_{20} \text{ at (A)}$$

But by KCL at (B), current in  $100 \Omega$  is also  $i_{sh} + i_{20} = i_y$ .

$$\text{Then at (C): } -i_y - 5i_y - 1.5 \text{ A} = 0$$

$$\Rightarrow i_y = -\frac{1.5}{6} = -0.25 \text{ A}$$

Room for extra work

Now the currents labeled  $i_1$  and  $i_2$  are 0, so KVL around the outside gives

$$\begin{aligned} 15V_x + 0 + 40 + 10 \cdot i_y - V_x - 20 \\ - 150 i_y - 6 i_y \cdot 10 = 0 \end{aligned}$$

$$\therefore 14V_x - 200 i_y = -20$$

$$\Rightarrow V_x = \frac{-20 + 200(-0.25)}{14}$$

$$\underline{V_x = -5.00 \text{ V}}$$

(not asked)

For  $V_3$ :

$$V_3 - 100 i_y - 150 i_y - 6 i_y \cdot 10 = 0$$

$$\underline{V_3 = 310 i_y = -77.5 \text{ V}}$$

KVL, KCL sign error -3

NV sign error -3

Correct eqns,  
wrong answer -3

leave out  $V$  across  
current source } -6  
→ missing eqn.