#### DO NOT BEGIN THIS EXAM UNTIL TOLD TO START

Name:	 		
PeopleSoft ID:			

# **ECE 3317**Applied Electromagnetic Waves

#### Exam II Nov. 28. 2018

- 1. This exam is open book and open notes. However, you are not allowed to use a computer or any electronic device other than a calculator. Any devices that may be used to communicate are not allowed.
- 2. Show all of your work. No credit will be given if the work required to obtain the solutions is not shown.
- 3. Perform all your work on the exam in the space allowed.
- 4. Write neatly. You will not be given credit for work that is not **easily** legible.
- 5. Leave answers in terms of the parameters given in the problem.
- 6. Show units in all of your final answers.
- 7. Circle your final answers.
- 8. Double-check your answers. For simpler problems, partial credit may not be given.
- 9. If you have any questions, ask the instructor. You will not be given credit for work that is based on a wrong assumption.
- 10. Make sure you sign the academic honesty statement on the next page.

## **Academic Honesty Statement**

I agree to abide by the UH Academic Honesty Policy during this exam. I understand that the punishment for violating this policy will be most severe, including getting an F in the class and getting expelled from the University.
Signature

#### **FORMULA SHEET**

$$\nabla \times \underline{\mathscr{E}} = -\frac{\partial \underline{\mathscr{B}}}{\partial t}$$

$$\nabla \times \mathcal{\underline{H}} = \underline{\mathcal{J}} + \frac{\partial \mathcal{\underline{Y}}}{\partial t}$$

$$\nabla \cdot \mathcal{B} = 0$$

$$\nabla \cdot \mathcal{Q} = \rho_{v}$$

$$\nabla \times \underline{E} = -j\omega \underline{B}$$

$$\nabla \times \underline{H} = \underline{J} + j\omega \underline{D}$$

$$\nabla \cdot \underline{B} = 0$$

$$\nabla \cdot \underline{D} = \rho_{v}$$

$$c = \frac{1}{\sqrt{\mu_0 \varepsilon_0}}$$

$$\varepsilon_0 \doteq 8.8541878 \times 10^{-12} \text{ [F/m]}$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ [H/m]}$$

$$\oint_{S} \left( \underline{\mathscr{E}} \times \underline{\mathscr{H}} \right) \cdot \hat{n} \, dS = -\int_{V} \sigma \left| \underline{\mathscr{E}} \right|^{2} \, dV - \int_{V} \frac{\partial}{\partial t} \left( \frac{1}{2} \, \mu \left| \underline{\mathscr{H}} \right|^{2} \right) dV - \int_{V} \frac{\partial}{\partial t} \left( \frac{1}{2} \, \varepsilon \left| \underline{\mathscr{E}} \right|^{2} \right) dV$$

$$\underline{\mathscr{G}} = \underline{\mathscr{E}} \times \underline{\mathscr{H}}$$

$$\underline{S} \equiv \frac{1}{2} \left( \underline{E} \times \underline{H}^* \right)$$

$$C = \frac{2\pi\varepsilon_0\varepsilon_r}{\ln\left(\frac{b}{a}\right)} \quad [F/m]$$

$$L = \frac{\mu_0}{2\pi} \ln\left(\frac{b}{a}\right) \quad [H/m]$$

$$G = \frac{2\pi\sigma_d}{\ln\left(\frac{b}{a}\right)} \quad [S/m]$$

$$R = \frac{1}{\sigma_m \delta} \left( \frac{1}{2\pi a} + \frac{1}{2\pi b} \right) \quad \left[ \Omega / \mathbf{m} \right]$$

$$\frac{\partial v}{\partial z} = -Ri - L\frac{\partial i}{\partial t}$$

$$\frac{\partial I}{\partial z}i = -Gv - C\frac{\partial v}{\partial t}$$

$$v(z,t) = f(z-c_{d}t) + g(z+c_{d}t)$$

$$i(z,t) = \frac{1}{Z_0} \left[ f(z - c_d t) - g(z + c_d t) \right]$$

$$v(z,t) = v_{g}(t - z / c_{d})$$

$$\Gamma_g = \left(\frac{R_g - Z_0}{R_g + Z_0}\right) \qquad \Gamma_L = \left(\frac{R_L - Z_0}{R_L + Z_0}\right)$$

$$V^{+} = \left(\frac{Z_0}{R_{\rho} + Z_0}\right) V_0$$

$$\Gamma_L(t) = 1 - 2e^{-(t-T)/\tau}, \quad t \ge T$$
  $\tau = Z_0 C_L$ 

$$\Gamma_L(t) = -1 + 2e^{-(t-T)/\tau}, \ t \ge T$$
  $\tau = L_L / Z_0$ 

$$V(z) = Ae^{-\gamma z} + Be^{+\gamma z}$$

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)}$$

$$LC = \mu \varepsilon = \frac{1}{c_d^2}$$

$$\gamma = \alpha + j\beta$$

$$k_z = -j\gamma = \beta - j\alpha$$

$$v_{p} = \frac{\omega}{\beta}$$

$$\beta = \frac{2\pi}{\lambda}$$

attenuation = 
$$\left(\frac{20}{\ln 10}\right)\alpha = (8.6859)\alpha$$
 [dB/m]

$$Z_0 = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$

$$i(z) = \left(\frac{1}{Z_0}\right) \left[Ae^{-\gamma z} - Be^{+\gamma z}\right]$$

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$V(z) = A\left(e^{-\gamma z} + \Gamma_L e^{+\gamma z}\right)$$

$$I(z) = \frac{1}{Z_0} A \left( e^{-\gamma z} - \Gamma_L e^{+\gamma z} \right)$$

$$Z_{in}(z) = \frac{V(z)}{I(z)}$$

$$Z_{in}(z) = Z_0 \left( \frac{1 + \Gamma_L e^{+2\gamma z}}{1 - \Gamma_L e^{+2\gamma z}} \right)$$

$$Z_{in} = Z_0 \left( \frac{Z_L + Z_0 \tanh(\gamma l)}{Z_0 + Z_L \tanh(\gamma l)} \right)$$

$$Z_{in} = Z_0 \left( \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)} \right)$$

$$Z_{in}(z) = Z_0 \left( \frac{1 + \Gamma(z)}{1 - \Gamma(z)} \right)$$

$$Z_{in} = jZ_0 \tan(\beta l)$$

$$Z_{in} = -jZ_0 \cot(\beta l)$$

$$\Gamma_L = |\Gamma_L| e^{j\phi}$$

$$|V(z)| = |A| \left| 1 + \left| \Gamma_{L} \right| e^{+j(\phi + 2\beta z)} \right|$$

$$\left| \frac{V(z)}{V^{+}} \right| = \left| 1 + \left| \Gamma_{L} \right| e^{+j(\phi + 2\beta z)} \right| = \left| 1 + \Gamma(z) \right|$$

$$\Gamma(z) = \Gamma_L e^{j2\beta z} = \Gamma_L e^{-j2\beta l}$$

$$VSWR \equiv \frac{V_{\text{max}}}{V}$$

$$VSWR \equiv \frac{1 + \left| \Gamma_{L} \right|}{1 - \left| \Gamma_{L} \right|}$$

$$SWR = \max\left(\frac{R_L}{Z_0}, \frac{Z_0}{R_L}\right)$$

$$Z_{in}^{N}(z) = \frac{1 + \Gamma_{L} e^{+2j\beta z}}{1 - \Gamma_{L} e^{+2j\beta z}}$$

$$Z_{in} = \frac{Z_{0T}^2}{Z_L}$$

$$Z_{0T} = \sqrt{Z_0 R_L}$$

$$k = \omega \sqrt{\mu \varepsilon}$$

$$\eta = \sqrt{\frac{\mu}{\varepsilon}}$$

$$\eta_0 = \sqrt{\frac{\mu_0}{\varepsilon_0}} \doteq 376.730313 \left[\Omega\right]$$

$$\underline{S} = \hat{\underline{z}} \frac{\left| E_0 \right|^2}{2\eta}$$

$$v_p = \frac{\omega}{k}$$

$$\lambda = \frac{2\pi}{k}$$

$$\lambda = \frac{c_d}{f}$$

$$E_x = E_0 e^{-jkz}$$

$$H_{y} = \frac{1}{\eta} E_0 e^{-jkz}$$

$$\varepsilon_c = \varepsilon - j \left( \frac{\sigma}{\omega} \right)$$

$$k = k' - jk''$$

$$\lambda = \frac{2\pi}{k'}$$

$$d_p = 1/k''$$

$$\tan \delta = \frac{\varepsilon_c''}{\varepsilon_c'} = \frac{\sigma}{\omega \varepsilon}$$

$$\delta = d_p = \sqrt{\frac{2}{\omega\mu\sigma}}$$

$$k' \approx k'' \approx \frac{1}{\delta}$$

$$Z_s \equiv \frac{E_{x0}}{J_{sx}}$$

$$R_{s} = \frac{1}{\sigma \delta} = \sqrt{\frac{\omega \mu}{2\sigma}}$$

$$X_s = R_s$$

$$R = X = R_s \left(\frac{l}{2\pi a}\right)$$

$$R = R_s \left( \frac{1}{2\pi a} + \frac{1}{2\pi b} \right)$$

$$\underline{E}(z) = (\hat{\underline{x}} E_x + \hat{\underline{y}} E_y) e^{-jkz}$$

(a) 
$$0 < \beta < \pi$$
 LHEP

(b) 
$$-\pi < \beta < 0$$
 RHEP

$$\gamma = \tan^{-1} \left( \frac{b}{a} \right)$$

$$0 \le \gamma \le 90^{\circ}$$

$$AR = \left|\cot \xi\right|$$

$$\xi > 0$$
: LHEP

$$\xi < 0$$
: RHEP

where

$$\sin 2\xi = \sin 2\gamma \sin \beta$$

$$-45^{\circ} \le \xi \le +45^{\circ}$$

$$n_1 \sin \theta_i = n_2 \sin \theta_i$$

$$n_1 \sin \theta_c = n_2 \sin 90^\circ$$

$$\theta_i = \theta_b = \tan^{-1} \sqrt{\frac{\varepsilon_2}{\varepsilon_1}}$$

$$\Gamma_{TM} = \frac{Z_2^{TM} - Z_1^{TM}}{Z_2^{TM} + Z_1^{TM}}$$

$$\Gamma_{TE} = \frac{Z_2^{TE} - Z_1^{TE}}{Z_2^{TE} + Z_1^{TE}}$$

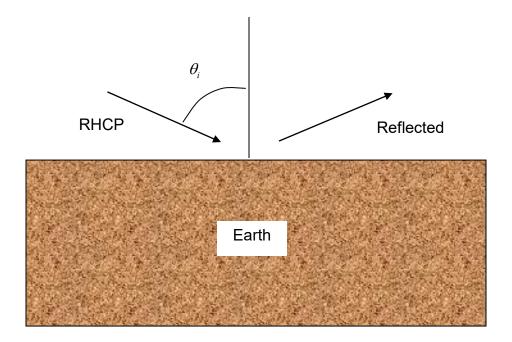
$$Z_1^{TE} = \left(\frac{\omega \mu_1}{k_{zi}}\right) \qquad Z_2^{TE} = \left(\frac{\omega \mu_2}{k_{zt}}\right)$$

$$Z_1^{TM} = \left(\frac{k_{zi}}{\omega \varepsilon_1}\right)$$
  $Z_2^{TM} = \left(\frac{k_{zt}}{\omega \varepsilon_2}\right)$ 

#### Problem 1 (30 pts)

A RHCP plane wave at 1.575 GHz from a GPS satellite is incident on the earth. The earth has a relative permittivity of  $\varepsilon_r = 6.0$  and is taken as lossless. The angle of incidence is  $\theta_i = 60^{\circ}$ . The power density in the incident plane wave from the GPS satellite (at an altitude of 20,200 km) is  $10^{-13}$  [W/m<sup>2</sup>].

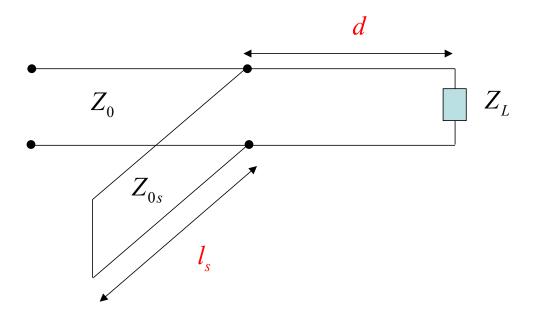
- (a) Calculate the power density in the reflected plane wave.
- (b) Calculate the percentage of the reflected power density that is in the  $TM_z$  polarization.
- (c) What would the angle of incidence have to be if we wanted no power to be reflected in the  $TM_z$  polarization?



#### Problem 2 (40 pts)

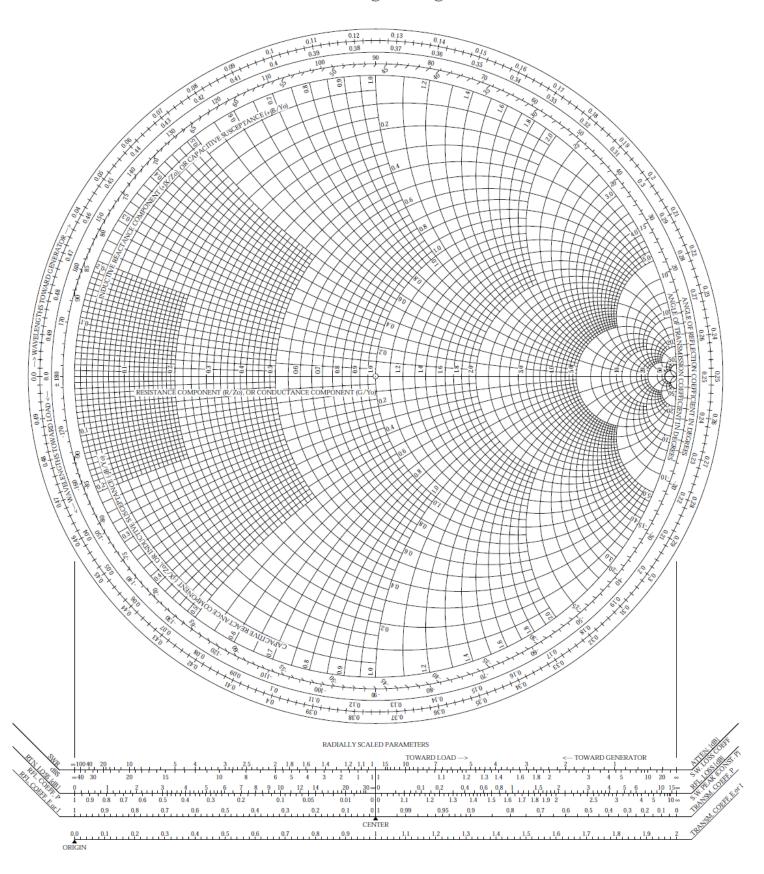
A twin-lead transmission line with  $Z_0 = 300 \ [\Omega]$  is connected to a load impedance that is given by  $Z_L = 100 - j(100) \ [\Omega]$ . Assume that the frequency is 500 MHz and the effective permittivity of the line is 1.0 (the line is in air). In order to match the line to the load, a shorted-circuited stub with characteristic impedance  $Z_{0s} = Z_0$  is inserted a distance d from the load.

Determine the distance d using the Smith chart that is on the next page (use the smallest d possible). Then find the length  $l_s$  of the stub line using the Smith chart that is attached.



## The Complete Smith Chart

Black Magic Design



#### Problem 3 (30 pts)

Consider the following plane wave that is traveling in air:

$$\underline{E} = \left[ (1 - j3) \underline{\hat{y}} + (2 + j) \underline{\hat{z}} \right] e^{-jk_0 x}.$$

- (a) Find the polarization (linear, circular, or elliptical) and handedness (left-handed or right-handed) for the wave.
- (b) Find the axial ratio of this wave.
- (c) Find the magnetic field vector for this plane wave.