#### DO NOT BEGIN THIS EXAM UNTIL TOLD TO START

Name:		
PeopleSoft ID:		

# ECE 3317 Applied Electromagnetic Waves FINAL EXAM May 7, 2013

- 1. This exam is open book and open notes. However, you are not allowed to use a computer or any electronic device other than a calculator. Any devices that may be used to communicate are not allowed.
- 2. Show all of your work. No credit will be given if the work required to obtain the solutions is not shown.
- 3. Perform all your work on the exam in the space allowed.
- 4. Write neatly. You will not be given credit for work that is not **easily** legible.
- 5. Leave answers in terms of the parameters given in the problem.
- 6. Show units in all of your final answers.
- 7. Circle your final answers.
- 8. Double-check your answers. For simpler problems, partial credit may not be given.
- 9. If you have any questions, ask the instructor. You will not be given credit for work that is based on a wrong assumption.
- 10. Make sure you sign the academic honesty statement on the next page.

## **Academic Honesty Statement**

I agree to abide by the UH Academic Honesty Policy during this exam. I understand that the punishment for violating this policy will be most severe, including getting an F in the class and getting expelled from the University.
Signature

$$\nabla \times \underline{\mathscr{E}} = -\frac{\partial \underline{\mathscr{B}}}{\partial t}$$

$$\nabla \times \mathcal{\underline{H}} = \underline{\mathcal{J}} + \frac{\partial \mathcal{\underline{Y}}}{\partial t}$$

$$\nabla \cdot \mathcal{B} = 0$$

$$\nabla \cdot \mathcal{Q} = \rho_{v}$$

$$\nabla \times \underline{E} = -j\omega \underline{B}$$

$$\nabla \times \underline{H} = \underline{J} + j\omega \underline{D}$$

$$\nabla \cdot \underline{B} = 0$$

$$\nabla \cdot \underline{D} = \rho_{v}$$

$$c = \frac{1}{\sqrt{\mu_0 \mathcal{E}_0}}$$

$$\varepsilon_0 \doteq 8.8541878 \times 10^{-12} \text{ [F/m]}$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ [H/m]}$$

$$\oint_{S} \left( \underline{\mathscr{E}} \times \underline{\mathscr{H}} \right) \cdot \hat{n} \, dS = -\int_{V} \sigma \left| \underline{\mathscr{E}} \right|^{2} \, dV - \int_{V} \frac{\partial}{\partial t} \left( \frac{1}{2} \mu \left| \underline{\mathscr{H}} \right|^{2} \right) dV - \int_{V} \frac{\partial}{\partial t} \left( \frac{1}{2} \varepsilon \left| \underline{\mathscr{E}} \right|^{2} \right) dV$$

$$\underline{\mathscr{S}} = \underline{\mathscr{E}} \times \underline{\mathscr{H}}$$

$$\underline{S} = \frac{1}{2} \left( \underline{E} \times \underline{H}^* \right)$$

$$C = \frac{2\pi\varepsilon_0\varepsilon_r}{\ln\left(\frac{b}{a}\right)} \quad [F/m]$$

$$L = \frac{\mu_0}{2\pi} \ln \left( \frac{b}{a} \right) \quad [H/m]$$

$$G = \frac{2\pi\sigma_d}{\ln\left(\frac{b}{a}\right)} \quad [S/m]$$

$$R = \frac{1}{\sigma_m \delta} \left( \frac{1}{2\pi a} + \frac{1}{2\pi b} \right) \left[ \Omega/m \right]$$

$$\frac{\partial v}{\partial z} = -Ri - L\frac{\partial i}{\partial t}$$

$$\frac{\partial I}{\partial z}i = -Gv - C\frac{\partial v}{\partial t}$$

$$v(z,t) = f(z-c_{t}t) + g(z+c_{t}t)$$

$$i(z,t) = \frac{1}{Z_0} \left[ f(z - c_d t) - g(z + c_d t) \right]$$

$$v(z,t) = v_g(t - z/c_d)$$

$$\Gamma_g = \left(\frac{R_g - Z_0}{R_g + Z_0}\right) \qquad \Gamma_L = \left(\frac{R_L - Z_0}{R_L + Z_0}\right)$$

$$V^{+} = \left(\frac{Z_0}{R_g + Z_0}\right) V_0$$

$$\Gamma_L(t) = 1 - 2e^{-(t-T)/\tau}, \quad t \ge T$$
  $\tau = Z_0C_L$ 

$$\Gamma_L(t) = -1 + 2e^{-(t-T)/\tau}, \ t \ge T$$
  $\tau = L_L / Z_0$ 

$$V(z) = Ae^{-\gamma z} + Be^{+\gamma z}$$

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)}$$

$$LC = \mu \varepsilon = \frac{1}{c_d^2}$$

$$\gamma = \alpha + j\beta$$

$$k_{z} = -j\gamma = \beta - j\alpha$$

$$v_{p} = \frac{\omega}{\beta}$$

$$\beta = \frac{2\pi}{\lambda}$$

attenuation = 
$$\left(\frac{20}{\ln 10}\right)\alpha = (8.6859)\alpha$$
 [dB/m]

$$Z_0 = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$

$$i(z) = \left(\frac{1}{Z_0}\right) \left[Ae^{-\gamma z} - Be^{+\gamma z}\right]$$

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$V(z) = A\left(e^{-\gamma z} + \Gamma_L e^{+\gamma z}\right)$$

$$I(z) = \frac{1}{Z_0} A \left( e^{-\gamma z} - \Gamma_L e^{+\gamma z} \right)$$

$$Z_{in}(z) = \frac{V(z)}{I(z)}$$

$$Z_{in}(z) = Z_0 \left( \frac{1 + \Gamma_L e^{+2\gamma z}}{1 - \Gamma_L e^{+2\gamma z}} \right)$$

$$Z_{in} = Z_0 \left( \frac{Z_L + Z_0 \tanh(\gamma l)}{Z_0 + Z_L \tanh(\gamma l)} \right)$$

$$Z_{in} = Z_0 \left( \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)} \right)$$

$$Z_{in}(z) = Z_0 \left( \frac{1 + \Gamma(z)}{1 - \Gamma(z)} \right)$$

$$Z_{in} = jZ_0 \tan(\beta l)$$

$$Z_{in} = -jZ_0 \cot(\beta l)$$

$$\Gamma_L = |\Gamma_L| e^{j\phi}$$

$$|V(z)| = |A| |1 + |\Gamma_{L}| e^{+j(\phi+2\beta z)}|$$

$$\left| \frac{V(z)}{V^{+}} \right| = \left| 1 + \left| \Gamma_{L} \right| e^{+j(\phi+2\beta z)} \right| = \left| 1 + \Gamma(z) \right|$$

$$\Gamma(z) = \Gamma_L e^{j2\beta z} = \Gamma_L e^{-j2\beta l}$$

$$VSWR \equiv \frac{V_{\text{max}}}{V_{\text{min}}}$$

$$VSWR = \frac{1 + \left| \Gamma_{L} \right|}{1 - \left| \Gamma_{L} \right|}$$

$$SWR = \max\left(\frac{R_L}{Z_0}, \frac{Z_0}{R_L}\right)$$

$$Z_{in}^{N}(z) = \frac{1 + \Gamma_{L} e^{+2j\beta z}}{1 - \Gamma_{L} e^{+2j\beta z}}$$

$$Z_{in} = \frac{Z_{0T}^2}{Z_I}$$

$$Z_{0T} = \sqrt{Z_0 R_L}$$

$$k = \omega \sqrt{\mu \varepsilon}$$

$$\eta = \sqrt{\frac{\mu}{\varepsilon}}$$

$$\eta_0 = \sqrt{\frac{\mu_0}{\varepsilon_0}} \doteq 376.730313 \, [\Omega]$$

$$\underline{S} = \hat{\underline{z}} \frac{\left| E_0 \right|^2}{2\eta}$$

$$v_p = \frac{\omega}{k}$$

$$\lambda = \frac{2\pi}{k}$$

$$\lambda = \frac{c_d}{f}$$

$$E_x = E_0 e^{-jkz}$$

$$H_{y} = \frac{1}{\eta} E_{0} e^{-jkz}$$

$$\varepsilon_c = \varepsilon - j \left( \frac{\sigma}{\omega} \right)$$

$$k = k' - jk''$$

$$\lambda = \frac{2\pi}{k'}$$

$$d_p = 1/k''$$

$$\tan \delta = \frac{\varepsilon_c''}{\varepsilon_c'} = \frac{\sigma}{\omega \varepsilon}$$

$$\delta = d_p = \sqrt{\frac{2}{\omega\mu\sigma}}$$

$$k' \approx k'' \approx \frac{1}{\delta}$$

$$Z_{s} \equiv \frac{E_{x0}}{J_{sx}}$$

$$R_s = \frac{1}{\sigma \delta} = \sqrt{\frac{\omega \mu}{2\sigma}}$$

$$X_s = R_s$$

$$R = X = R_s \left( \frac{l}{2\pi a} \right)$$

$$R = R_s \left( \frac{1}{2\pi a} + \frac{1}{2\pi b} \right)$$

$$\underline{E}(z) = (\hat{\underline{x}} E_x + \hat{\underline{y}} E_y) e^{-jkz}$$

(a) 
$$0 < \beta < \pi$$
 LHEP

(b) 
$$-\pi < \beta < 0$$
 RHEP

$$n_1 \sin \theta_i = n_2 \sin \theta_t$$

$$\Gamma_{TM} = \frac{Z_2^{TM} - Z_1^{TM}}{Z_2^{TM} + Z_1^{TM}}$$

$$\Gamma_{TE} = \frac{Z_2^{TE} - Z_1^{TE}}{Z_2^{TE} + Z_1^{TE}}$$

$$Z_1^{TE} = \left(\frac{\omega \mu_1}{k_{zi}}\right) \qquad Z_2^{TE} = \left(\frac{\omega \mu_2}{k_{zt}}\right)$$

$$Z_1^{TM} = \left(\frac{k_{zi}}{\omega \varepsilon_1}\right)$$
  $Z_2^{TM} = \left(\frac{k_{zt}}{\omega \varepsilon_2}\right)$ 

$$\theta_i = \theta_b = \tan^{-1} \sqrt{\frac{\varepsilon_2}{\varepsilon_1}}$$

$$k_z = k\sqrt{1 - \left(f_c / f\right)^2}$$

$$f_c = \frac{c_d}{2\pi} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

$$\beta = k\sqrt{1-(f_c/f)^2}, \qquad f > f_c$$

$$\alpha = k_c \sqrt{1 - (f/f_c)^2}, \quad f < f_c$$

$$H_z(x, y, z) = A_{10} \cos\left(\frac{\pi x}{a}\right) e^{-jk_z z}$$

$$\lambda_g = \frac{\lambda}{\sqrt{1 - \left(f_c / f\right)^2}}$$

$$v_p = \frac{c_d}{\sqrt{1 - \left(f_c / f\right)^2}}$$

$$v_{g} = \frac{d\omega}{d\beta} = \frac{1}{\frac{d\beta}{d\omega}}$$

$$v_g = c_d \sqrt{1 - \left(f_c / f\right)^2}$$

$$\underline{S} = \hat{\underline{r}} \left( \frac{\left| \underline{\underline{E}} \right|^2}{2\eta_0} \right)$$

$$\underline{E}(r,\theta,\phi) = \left(\frac{e^{-jk_0r}}{r}\right)\underline{E}^F(\theta,\phi)$$

$$P_{rad} = \frac{1}{2\eta_0} \int_{0}^{2\pi} \int_{0}^{\pi} \left| \underline{E}^F (\theta, \phi) \right|^2 \sin \theta \, d\theta d\phi$$

$$D(\theta,\phi) = \frac{S_r(\theta,\phi)}{P_{rad}/(4\pi r^2)} \qquad r \to \infty$$

$$D(\theta,\phi) = \frac{4\pi \left| \underline{E}^{F}(\theta,\phi) \right|^{2}}{\int_{0}^{2\pi} \int_{0}^{\pi} \left| \underline{E}^{F}(\theta,\phi) \right|^{2} \sin \theta \, d\theta d\phi}$$

$$G(\theta,\phi) \equiv e_r D(\theta,\phi)$$

$$E_{\theta}^{F} = \frac{Il}{4\pi} (j\omega\mu_{0}) \sin\theta$$

$$H_{\phi}^{F} = \frac{Il}{4\pi} (jk_0) \sin \theta$$

$$D(\theta,\phi) = \frac{3}{2}\sin^2\theta$$

$$AF(\theta) \equiv \int_{-h}^{h} I(z') e^{+jk_0 z' \cos \theta} dz'$$

$$E_{\theta} \approx \frac{1}{4\pi} (j\omega\mu_0) \sin\theta \left(\frac{e^{-jk_0r}}{r}\right) AF(\theta)$$

$$AF(\theta) = 2\left(\frac{I_0}{\sin(k_0 h)}\right)\left[\frac{\cos(k_0 h \cos \theta) - \cos(k_0 h)}{k_0 \sin^2 \theta}\right]$$

$$D(\pi/2,\phi)=1.643$$

$$R_{rad} \approx 73 \ \left[\Omega\right]$$

$$V_{\mathit{Th}} = l_{\mathit{eff}} \left( \hat{\underline{l}} \cdot \underline{E}^{\mathit{inc}} \right)$$

$$Z_{Th} = Z_{in}$$

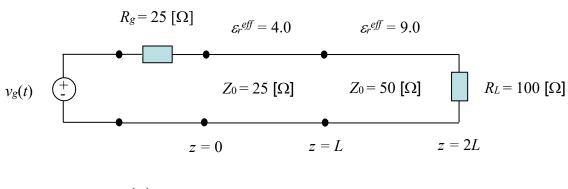
$$V^{monopole} = \frac{1}{2} V^{dipole}$$

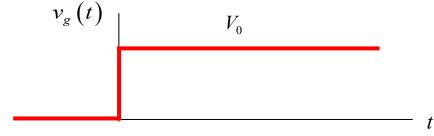
$$l_{\it eff} = \begin{cases} l/2, & l << \lambda_0 \text{ (electrically small)} \\ l\left(\frac{2}{\pi}\right) \cos\left(\frac{\pi}{2}\cos\theta\right) / \sin^2\theta, & l = \lambda_0/2 \text{ (resonant)} \end{cases}$$

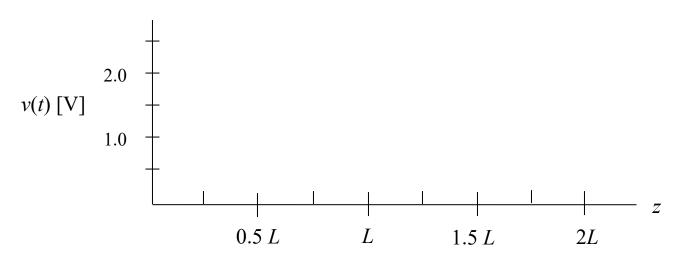
#### Problem 1 (40 pts.)

A step function of amplitude  $V_0 = 2.0$  [V] is applied at the input to the transmission line circuit shown below. Each of the two transmission lines has a length L that is 3.0 [m]. Each line has a different effective relative permittivity, however.

- a) Construct a bounce diagram that extends to a time of 50 [ns], labeling with 10 [ns] divisions on your time scale. (Make your bounce diagram on the next page.)
- b) Make an accurate "snapshot" trace of the voltage on the line at t = 35 [ns]. Make your snapshot trace on the graph shown below, plotting out to 2L. Label all voltage values on your plot as well as all z values at which the voltage on your plot changes. Also, indicate in which direction (to the left or right) each "wavefront" (point of voltage discontinuity) is moving.







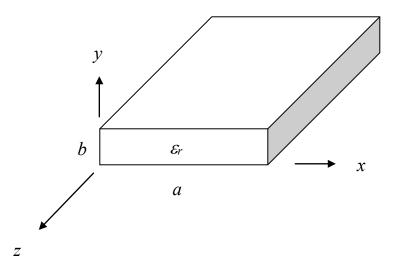
#### Problem 2 (40 pts.)

A rectangular waveguide has a dominant TE<sub>10</sub> mode that has an electric field that is given by

$$\underline{E} = \underline{\hat{y}} \sin\left(\frac{\pi x}{a}\right) e^{-jk_z z}.$$

The waveguide is filled with a lossless material having a relative permittivity  $\varepsilon_r$ .

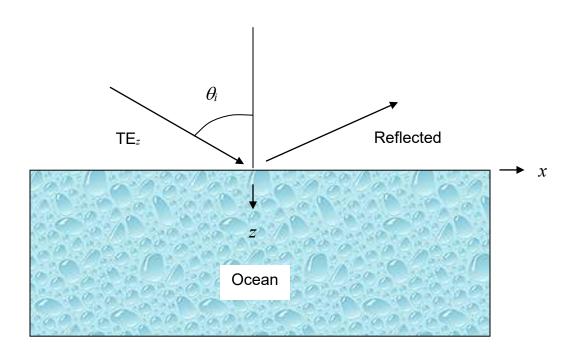
- a) Find the complex power flow  $P_f$  that is flowing in the z direction down the waveguide at any frequency f. (Evaluate all integrals so that your result is a closed-form expression.)
- b) Find the time-average power flowing down the waveguide when  $f > f_c$ . Your result should be in closed form. Do not leave any "Re" or "Im" operators in your answer.
- c) Find the VARS flowing down the waveguide when  $f < f_c$ . Your result should be in closed form. Do not leave any "Re" or "Im" operators in your answer.



### Problem 3 (40 pts.)

A TE<sub>z</sub> plane wave at a frequency of 1.0 GHz is incident on the ocean (which is nonmagnetic). The ocean has a relative permittivity of  $\varepsilon_r = 81$  and a conductivity of  $\sigma = 4.0$  [S/m]. The angle of incidence is  $\theta_i = 60^{\circ}$ .

- a) Calculate the total power density in the reflected plane wave relative to that of the incident plane wave.
- b) Find the distance z below the surface of the ocean for which the field  $E_y$  has attenuated by 10 dB from what it is at the surface (z = 0).



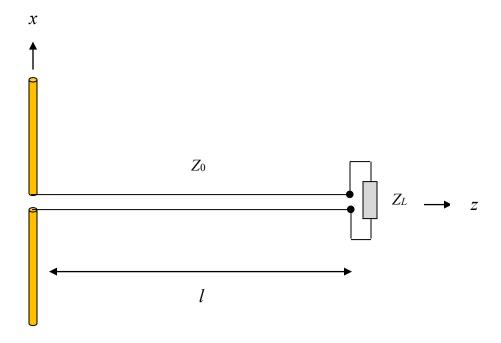
#### Problem 4 (60 pts.)

A plane wave from a distant transmitter is incident on a receive antenna as shown below. The incident wave is a RHCP wave and it has the mathematical form

$$\underline{E} = \left(\underline{\hat{x}}(1) + \underline{\hat{y}}(-j)\right) e^{-jk_0 z}.$$

The frequency is 1.0 [GHz]. The receive antenna is a resonant half-wavelength dipole wire antenna that is aligned in the x direction. The receive antenna is connected to a lossless transmission line of length l, which is then connected to a load. The load is an impedance  $Z_L = 100 - j(50)$  [ $\Omega$ ] while the transmission line has a characteristic impedance of  $Z_0 = 50$  [ $\Omega$ ]. The length of the transmission line is  $l = \lambda / 8$ , where  $\lambda$  is the wavelength on the transmission line.

- a) Find the Thévenin equivalent circuit for the receive antenna.
- b) Find the input impedance seen by the receive antenna, using the Smith chart (attached at the end of the exam).
- c) Find the power being delivered to the load.



#### Problem 5 (50 pts.)

An antenna radiates an electric field in the far field that is given by

$$\underline{E}^{F}(\theta,\phi) = \begin{cases} E_{0}(\hat{\theta}\cos\theta\cos\phi + \hat{\phi}\sin\phi), & \theta \leq \pi/2\\ \underline{0}, & \theta \geq \pi/2 \end{cases}$$

where  $E_0$  is a constant.

- a) Calculate the directivity  $D(\theta, \phi)$ . Make sure that you evaluate all integrals to get a closed-form result.
- b) Find the direction ( $\theta_{max}$ ,  $\phi_{max}$ ) that maximizes the directivity, and find the value of the maximum directivity; i.e., find  $D_{max} = D(\theta_{max}, \phi_{max})$ .
- c) If the antenna radiates 1 [W] of power, find the magnitude of the electric field vector in the direction  $\theta = 0$ , at a distance of 1.0 [km] from the antenna.

