DO NOT BEGIN THIS EXAM UNTIL TOLD TO START

Name:	SOLUTION	
PeopleSoft ID:		

ECE 3317 Applied Electromagnetic Waves FINAL EXAM Dec. 12, 2018

- 1. This exam is open book and open notes. However, you are not allowed to use a computer or any electronic device other than a calculator. Any devices that may be used to communicate are not allowed.
- 2. Show all of your work. No credit will be given if the work required to obtain the solutions is not shown.
- 3. Perform all your work on the exam in the space allowed.
- 4. Write neatly. You will not be given credit for work that is not **easily** legible.
- 5. Leave answers in terms of the parameters given in the problem.
- 6. Show units in all of your final answers.
- 7. Circle your final answers.
- 8. Double-check your answers. For simpler problems, partial credit may not be given.
- 9. If you have any questions, ask the instructor. You will not be given credit for work that is based on a wrong assumption.
- 10. Make sure you sign the academic honesty statement on the next page.

Academic Honesty Statement

I agree to abide by the UH Academic Honesty Policy during this exam. I understand that the punishment for violating this policy will be most severe, including getting an F in the class and getting expelled from the University.
Signature

$$\nabla \times \underline{\mathscr{E}} = -\frac{\partial \underline{\mathscr{B}}}{\partial t}$$

$$\nabla \times \underline{\mathcal{H}} = \underline{\mathcal{J}} + \frac{\partial \underline{\mathcal{D}}}{\partial t}$$

$$\nabla \cdot \mathcal{B} = 0$$

$$\nabla \cdot \mathcal{Q} = \rho_{v}$$

$$\nabla \times E = -j\omega B$$

$$\nabla \times \underline{H} = \underline{J} + j\omega \underline{D}$$

$$\nabla \cdot B = 0$$

$$\nabla \cdot \underline{D} = \rho_{v}$$

$$c = \frac{1}{\sqrt{\mu_0 \varepsilon_0}}$$

$$\varepsilon_0 \doteq 8.8541878 \times 10^{-12} \text{ [F/m]}$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ [H/m]}$$

$$\oint_{S} \left(\underline{\mathscr{E}} \times \underline{\mathscr{H}} \right) \cdot \hat{n} \, dS = -\int_{V} \sigma \left| \underline{\mathscr{E}} \right|^{2} \, dV - \int_{V} \frac{\partial}{\partial t} \left(\frac{1}{2} \, \mu \left| \underline{\mathscr{H}} \right|^{2} \right) dV - \int_{V} \frac{\partial}{\partial t} \left(\frac{1}{2} \, \varepsilon \left| \underline{\mathscr{E}} \right|^{2} \right) dV$$

$$\underline{\mathscr{S}} = \underline{\mathscr{E}} \times \underline{\mathscr{H}}$$

$$\underline{S} \equiv \frac{1}{2} \left(\underline{E} \times \underline{H}^* \right)$$

$$C = \frac{2\pi\varepsilon_0\varepsilon_r}{\ln\left(\frac{b}{a}\right)} \quad [F/m]$$

$$L = \frac{\mu_0}{2\pi} \ln\left(\frac{b}{a}\right) \quad [H/m]$$

$$G = \frac{2\pi\sigma_d}{\ln\left(\frac{b}{a}\right)} \quad \left[\text{S/m}\right]$$

$$R = \frac{1}{\sigma_m \delta} \left(\frac{1}{2\pi a} + \frac{1}{2\pi b} \right) \left[\Omega / m \right]$$

$$\frac{\partial v}{\partial z} = -Ri - L\frac{\partial i}{\partial t}$$

$$\frac{\partial I}{\partial z}i = -Gv - C\frac{\partial v}{\partial t}$$

$$v(z,t) = f(z-c_{d}t) + g(z+c_{d}t)$$

$$i(z,t) = \frac{1}{Z_0} \left[f(z - c_d t) - g(z + c_d t) \right]$$

$$v(z,t) = v_g(t - z / c_d)$$

$$\Gamma_g = \left(\frac{R_g - Z_0}{R_g + Z_0}\right) \qquad \Gamma_L = \left(\frac{R_L - Z_0}{R_L + Z_0}\right)$$

$$\Gamma_L = \left(\frac{R_L - Z_0}{R_L + Z_0}\right)$$

$$V^{+} = \left(\frac{Z_0}{R_g + Z_0}\right) V_0$$

$$\Gamma_L(t) = 1 - 2e^{-(t-T)/\tau}, \quad t \ge T$$
 $\tau = Z_0C_L$

$$\tau = Z_0 C_L$$

$$\Gamma_L(t) = -1 + 2e^{-(t-T)/\tau}, \ t \ge T$$
 $\tau = L_L / Z_0$

$$\tau = L_L / Z_0$$

$$V(z) = Ae^{-\gamma z} + Be^{+\gamma z}$$

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)}$$

$$LC = \mu \varepsilon = \frac{1}{c_d^2}$$

$$\gamma = \alpha + j\beta$$

$$k_z = -j\gamma = \beta - j\alpha$$

$$v_{p} = \frac{\omega}{\beta}$$

$$\beta = \frac{2\pi}{\lambda}$$

attenuation =
$$\left(\frac{20}{\ln 10}\right)\alpha = (8.6859)\alpha$$
 [dB/m]

$$Z_0 = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$

$$i(z) = \left(\frac{1}{Z_0}\right) \left[Ae^{-\gamma z} - Be^{+\gamma z}\right]$$

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$V(z) = A\left(e^{-\gamma z} + \Gamma_L e^{+\gamma z}\right)$$

$$I(z) = \frac{1}{Z_0} A \left(e^{-\gamma z} - \Gamma_L e^{+\gamma z} \right)$$

$$Z_{in}(z) = \frac{V(z)}{I(z)}$$

$$Z_{in}(z) = Z_0 \left(\frac{1 + \Gamma_L e^{+2\gamma z}}{1 - \Gamma_L e^{+2\gamma z}} \right)$$

$$Z_{in} = Z_0 \left(\frac{Z_L + Z_0 \tanh(\gamma l)}{Z_0 + Z_L \tanh(\gamma l)} \right)$$

$$Z_{in} = Z_0 \left(\frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)} \right)$$

$$Z_{in}(z) = Z_0 \left(\frac{1 + \Gamma(z)}{1 - \Gamma(z)} \right)$$

$$Z_{in} = jZ_0 \tan(\beta l)$$

$$Z_{in} = -jZ_0 \cot(\beta l)$$

$$\Gamma_L = \left| \Gamma_L \right| e^{j\phi}$$

$$|V(z)| = |A| |1 + |\Gamma_{L}| e^{+j(\phi+2\beta z)}|$$

$$\left| \frac{V(z)}{V^{+}} \right| = \left| 1 + \left| \Gamma_{L} \right| e^{+j(\phi+2\beta z)} \right| = \left| 1 + \Gamma(z) \right|$$

$$\Gamma(z) = \Gamma_L e^{j2\beta z} = \Gamma_L e^{-j2\beta l}$$

$$VSWR \equiv \frac{V_{\text{max}}}{V_{\text{min}}}$$

$$VSWR = \frac{1 + \left| \Gamma_{L} \right|}{1 - \left| \Gamma_{L} \right|}$$

$$SWR = \max\left(\frac{R_L}{Z_0}, \frac{Z_0}{R_L}\right)$$

$$Z_{in}^{N}(z) = \frac{1 + \Gamma_{L} e^{+2j\beta z}}{1 - \Gamma_{L} e^{+2j\beta z}}$$

$$Z_{in} = \frac{Z_{0T}^2}{Z_L}$$

$$Z_{0T} = \sqrt{Z_0 R_L}$$

$$k = \omega \sqrt{\mu \varepsilon}$$

$$\eta = \sqrt{\frac{\mu}{\varepsilon}}$$

$$\eta_0 = \sqrt{\frac{\mu_0}{\varepsilon_0}} \doteq 376.730313 \left[\Omega\right]$$

$$\underline{S} = \hat{\underline{z}} \frac{\left| E_0 \right|^2}{2\eta}$$

$$v_p = \frac{\omega}{k}$$

$$\lambda = \frac{2\pi}{k}$$

$$\lambda = \frac{c_d}{f}$$

$$E_x = E_0 e^{-jkz}$$

$$H_{y} = \frac{1}{n} E_0 e^{-jkz}$$

$$\varepsilon_c = \varepsilon - j \left(\frac{\sigma}{\omega} \right)$$

$$k = k' - jk''$$

$$\lambda = \frac{2\pi}{k'}$$

$$d_p = 1/k''$$

$$\tan \delta = \frac{\varepsilon_c''}{\varepsilon_c'} = \frac{\sigma}{\omega \varepsilon}$$

$$\delta = d_p = \sqrt{\frac{2}{\omega\mu\sigma}}$$

$$k' \approx k'' \approx \frac{1}{\delta}$$

$$Z_s \equiv \frac{E_{x0}}{J_{sx}}$$

$$R_s = \frac{1}{\sigma \delta} = \sqrt{\frac{\omega \mu}{2\sigma}}$$

$$X_s = R_s$$

$$R = X = R_s \left(\frac{l}{2\pi a} \right)$$

$$R = R_s \left(\frac{1}{2\pi a} + \frac{1}{2\pi b} \right)$$

$$\underline{E}(z) = (\hat{\underline{x}} E_x + \hat{\underline{y}} E_y) e^{-jkz}$$

(a)
$$0 < \beta < \pi$$
 LHEP

(b)
$$-\pi < \beta < 0$$
 RHEP

$$\gamma = \tan^{-1}\left(\frac{b}{a}\right)$$

$$0 \le \gamma \le 90^{\circ}$$

$$AR = \left| \cot \xi \right|$$

$$\xi > 0$$
: LHEP

$$\xi < 0$$
: RHEP

where

$$\sin 2\xi = \sin 2\gamma \sin \beta$$

$$-45^{\circ} \le \xi \le +45^{\circ}$$

$$n_1 \sin \theta_i = n_2 \sin \theta_t$$

$$n_1 \sin \theta_c = n_2 \sin 90^{\circ}$$

$$\theta_i = \theta_b = \tan^{-1} \sqrt{\frac{\varepsilon_2}{\varepsilon_1}}$$

$$Z_1^{TE} = \left(\frac{\omega \mu_1}{k_{zi}}\right)$$
 $Z_2^{TE} = \left(\frac{\omega \mu_2}{k_{zt}}\right)$

$$Z_1^{TM} = \left(\frac{k_{zi}}{\omega \varepsilon_1}\right)$$
 $Z_2^{TM} = \left(\frac{k_{zt}}{\omega \varepsilon_2}\right)$

$$k_z = k\sqrt{1 - \left(f_c / f\right)^2}$$

$$f_c = \frac{c_d}{2\pi} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

$$\beta = k\sqrt{1 - (f_c / f)^2}, \qquad f > f_c$$

$$\alpha = k_c \sqrt{1 - (f/f_c)^2}, \quad f < f_c$$

$$H_z(x, y, z) = A_{10} \cos\left(\frac{\pi x}{a}\right) e^{-jk_z z}$$

$$\lambda_g = \frac{\lambda}{\sqrt{1 - \left(f_c / f\right)^2}}$$

$$v_p = \frac{c_d}{\sqrt{1 - \left(f_c / f\right)^2}}$$

$$v_{g} = \frac{d\omega}{d\beta} = \frac{1}{\frac{d\beta}{d\omega}}$$

$$v_g = c_d \sqrt{1 - \left(f_c / f\right)^2}$$

$$\underline{S} = \hat{\underline{r}} \left(\frac{\left| \underline{\underline{E}} \right|^2}{2\eta_0} \right)$$

$$\underline{E}(r,\theta,\phi) = \left(\frac{e^{-jk_0r}}{r}\right)\underline{E}^F(\theta,\phi)$$

$$P_{rad} = \frac{1}{2\eta_0} \int_{0}^{2\pi} \int_{0}^{\pi} \left| \underline{E}^F (\theta, \phi) \right|^2 \sin \theta \, d\theta d\phi$$

$$D(\theta,\phi) = \frac{S_r(\theta,\phi)}{P_{rad}/(4\pi r^2)} \qquad r \to \infty$$

$$D(\theta,\phi) = \frac{4\pi \left| \underline{E}^{F}(\theta,\phi) \right|^{2}}{\int_{0}^{2\pi} \int_{0}^{\pi} \left| \underline{E}^{F}(\theta,\phi) \right|^{2} \sin \theta \, d\theta d\phi}$$

$$G(\theta,\phi) \equiv e_r D(\theta,\phi)$$

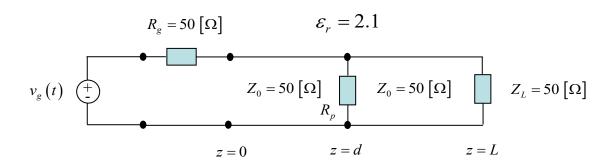
$$Z_{\mathit{Th}} = Z_{\mathit{in}}$$

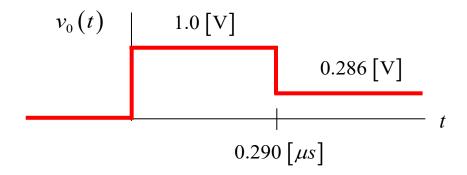
Problem 1 (20 pts.)

A Time Domain Reflectometer (TDR) is being used to determine a fault on a transmission line. The TDR has a voltage generator $v_g(t)$ that applies a voltage step function with an amplitude of $V_0 = 2.0$ [V] at the input to the transmission line circuit shown below. The transmission line has a relative permittivity of 2.1. The voltage waveform $v_0(t)$ that is recorded by the TDR at z = 0 is shown below. A partial short on the line (the fault) is modeled as a parallel resistance R_p as shown.

Determine the unknown resistance R_p and the distance d between the short and the TDR.

Support your answer by constructing a bounce diagram and using it to get the voltage $v_0(t)$ in terms of d and R_p .





$$d = 30 [m]$$

$$R_p = 10 [\Omega]$$

Problem 2 (20 pts.)

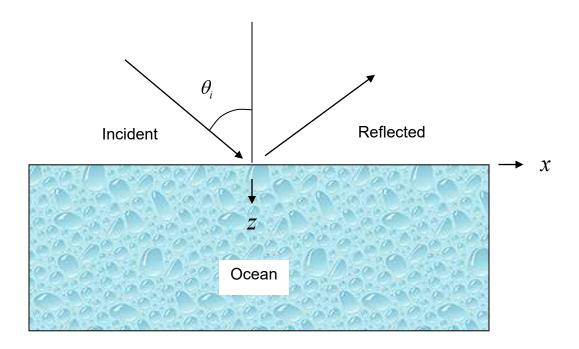
A 75 [Ω] RG-59 coaxial cable has an outer radius of b=1.85 [mm] and an inner radius of a=0.292 [mm]. The coax is filled with Teflon ($\varepsilon_r=2.25$) that has a loss tangent of 0.001. The conductors are made of copper, which is nonmagnetic and has a conductivity of $\sigma=3.0\times10^8$ [S/m]. Calculate the attenuation in dB when a 500 [MHz] signal travels along the cable a distance of 20 meters.

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R = 1.619 \, [\Omega/m]
G = 2.13 \times 10^{-4} \, [S/m]
L = 3.692 \times 10^{-7} \, [H/m]
C = 6.78 \times 10^{-11} \, [F/m]
\gamma = 0.0188 + j \, (15.72) \, [1/m]
\alpha = 0.0188 \, [np/m]
\alpha = 0.164 \, [dB/m]
Attenuation = 3.27 [dB]
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Problem 3 (20 pts.)

A TE_z plane wave at a frequency of 1.575 GHz (GPS frequency) is incident from the air onto the ocean (which is nonmagnetic). The ocean has a relative permittivity of $\varepsilon_r = 81$ and a conductivity of $\sigma = 4.0$ [S/m]. The angle of incidence is $\theta_i = 45^\circ$.

- a) Find percentage of power that is reflected from the ocean.
- b) Find the dB of attenuation in the plane wave as it travels from just below the surface down to a depth of 10 meters (z = 10 [m]).



$$\varepsilon_{rc} = 81.0 - j(45.65)$$

$$Z_1^{TE} = 532.77 \left[\Omega\right]$$

$$Z_2^{TE} = 37.86 + j(9.99) [\Omega]$$

$$\Gamma_{TE} = -0.867 + j(0.033)$$

$$P_r^{\%} = 75.2$$

$$k_{zt} = 307.0 - j(81.0) [1/m]$$

$$\alpha_{zt} = 81.0 \, [\text{np/m}]$$

$$\alpha_{zt} = 703.6 \, [dB/m]$$

Attenuation = 7036 [dB]

Problem 4 (20 pts.)

An air-filled rectangular waveguide is designed to transmit the dominant mode at an operating frequency of 8.0 [GHz]. The operating frequency should be halfway between the cutoff frequencies of the dominant mode and the next higher mode. Assume that b = a/2.

- a) Determine the dimensions a and b of the waveguide.
- b) Assume that we wish to have the dominant mode attenuated by at least 100 dB when the frequency is 5% below the cutoff frequency of the dominant mode. What is the minimum length of waveguide (in the z direction) required to do this?

```
a = 0.02811 \text{ [m]}

b = 0.01405 \text{ [m]}

\alpha = 34.90 \text{ [np/m]}

\alpha = 303.1 \text{ [dB/m]}

l = 0.330 \text{ [m]}
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Problem 5 (20 pts.)

NASA wants to use a large reflector antenna in the Deep Space Network (DSN) to communicate with a CubeSat satellite. A simple dipole antenna is used on the CubeSat. The frequency is 7.2 [GHz]. The transmit dish antenna in the DSN has a diameter of 34 [m]. Assume that the aperture efficiency of the dish is 70%.

- a) If the transmitted power is 100 [W], find the power density at the cubesat location, assuming that the cubesat is located near Mars, at a distance of 75 million [km].
- b) Find the power that the CubeSat antenna is able to deliver to a matched load, assuming ideal conditions (the polarization of the transmitted wave and the receive dipole antenna are perfectly aligned).

$$A_e^{dish} = 635.54 \left[\text{m}^2 \right]$$

$$G^{dish} = 4.607 \times 10^6$$

$$P_d = 6.52 \times 10^{-15} \, [W]$$

$$A_e^{dip} = 2.267 \times 10^{-4} \left[\text{m}^2 \right]$$

$$P_{rec} = 1.477 \times 10^{-18} \text{ [W]}$$

Bonus (20 pts.)

An antenna radiates an electric field in the far field that is given by

$$\underline{E}^{F}(\theta,\phi) = \begin{cases} E_{0}(\hat{\underline{\theta}}\cos\theta\cos\phi + \hat{\underline{\phi}}\sin\phi), & \theta \leq \pi/2\\ \underline{0}, & \theta \geq \pi/2, \end{cases}$$

where E_0 is a constant.

Calculate the directivity $D(\theta, \phi)$. Make sure that you evaluate all integrals to get a closed-form result.

$$D(\theta,\phi) = \frac{4\pi \left(\cos^2\theta\cos^2\phi + \sin^2\phi\right)}{\int\limits_0^{2\pi \pi/2} \left(\cos^2\theta\cos^2\phi + \sin^2\phi\right)\sin\theta \,d\theta d\phi}$$

$$D(\theta,\phi) = \frac{4\pi \left(\cos^2 \theta \cos^2 \phi + \sin^2 \phi\right)}{\pi \int_{0}^{\pi/2} \left(\cos^2 \theta + 1\right) \sin \theta \, d\theta}$$

$$D(\theta, \phi) = \frac{4(\cos^2\theta\cos^2\phi + \sin^2\phi)}{\int_0^{\pi/2} (\cos^2\theta + 1)\sin\theta \,d\theta}$$

$$D(\theta,\phi) = \frac{4(\cos^2\theta\cos^2\phi + \sin^2\phi)}{\int_0^{\pi/2} \cos^2\theta\sin\theta \,d\theta + \int_0^{\pi/2} \sin\theta \,d\theta}$$

$$D(\theta, \phi) = \frac{4(\cos^2\theta\cos^2\phi + \sin^2\phi)}{\frac{1}{3} + 1}$$

$$D(\theta, \phi) = 3(\cos^2\theta \cos^2\phi + \sin^2\phi)$$